DETERMINATION OF PROPERTIES OF AN OPTICAL DEVICE

BACKGROUND OF THE INVENTION

Field of the Invention

The present invention relates to the determination of properties of an optical device under test, e.g. the determination of the group delay of the optical

5   device. The group delay is a fundamental property of optical devices, such as single mode optical fibers or optical components such as Bragg gratings which devices are used in the optical transmission of information.

Discussion of the Background Art

As for the background of optical transmission of data it has to be said that the premier feature of optical fiber is its extremely low loss. This has made it the

10   dominant transmission medium for long link lengths. The loss characteristics of fiber determines were optical communication is practical. At 1550 nanometer single mode optical fiber has an attenuation of 0.2 dB/km. This allows fiber optic signals to be propagated through very long length of fiber without regeneration. Telecommunication systems use the 1300 and 1550 nm windows

15   for lowest loss in the fiber. Since a telecommunication system must cover a very large distance, the aforementioned attenuation of the single strength in the fiber is of high importance. Therefore, the loss characteristics of optical fiber often limit the distance that a signal can propagate in the fiber.

However, this is not always the case. In single mode fibers, chromatic

20   dispersion can limit the distance over which fiber optic signals can propagate. Chromatic dispersion describes the fact that the speed of signal propagation in the fiber depends on the wavelength of the light. The consequence is that as the signal propagates through a long length of fiber, the edges of the waveform of the signal start to become more rounded. Eventually, the adjacent bits start

25   to overlap in time causing the digital waveform to have poor readability. The amount of signal rounding depends beyond other parameters on the amount of chromatic dispersion in the cable. The problem of this pulse spreading is a

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problem in today's fiber communication because the increasing bit rate of the state of the art fiber communication systems, which bit rate reaches numbers of up to 40 Gbit/s per channel, brings the chromatic dispersion in the range of the bit resolution of such a 40 Gbit/s communication system.

5   To compensate chromatic dispersion in a fiber chirped Bragg gratings were developed. In these gratings different wavelengths of the pulse need different amounts of time traveling through the grating to compensate the dispersion caused by the fiber. The concept of this compensation is schematically illustrated in Fig. 1. Fig. 1 shows the original pulse 1 that is broadened to a

10   spreaded pulse 2 while traveling through a fiber 3. With the help of a circulator 4 a chirped Bragg grating 5 is introduced in the path of the pulse 2. As described before the resulting pulse 6 has the original shape of pulse 1 again.

However, to use such a component as a chirped Bragg grating it is necessary to have exact knowledge of the chromatic dispersion caused by such a grating.

15   The measurement of chromatic dispersion is accomplished by analyzing the group delay through the fiber as a function of wavelength. To measure the group delay a wavelength tunable optical source such as a tunable laser is used to generate coherent light at different wavelengths. In the prior art the wavelength of the tunable laser is then incremented step by step and for each

20   wavelength step a group delay is determined. Finally, the group delay details are used to calculate the chromatic dispersion coefficient. The disadvantage of measuring the group delays step by step is the time, which is needed to perform all the wavelength steps and the respective measurements. In other words this means it takes quite a long time to get data, which are so precise

25   that this precision is sufficient for the needs of the telecommunication industry.

<center>SUMMARY OF THE INVENTION</center>

Therefore, it is an object of the invention to provide an improved determination

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of a property of an optical device under test, e.g. the group delay of the optical device. The object is solved by the independent claims.

The main advantage of the present invention is that it is possible to measure the group delay and therefore the chromatic dispersion of an optical device, such as a fiber or a chirped Bragg grating, in a short amount of time while still keeping the precision of the data high enough for the needs of the telecommunication industry. This goal is reached by the invention by tuning the frequency of the coherent light beam of the laser from a maximum to a minimum of a given frequency range in a given time interval, e.g. making a wavelength sweep through the given wavelength range. Because of the inventive correction of effects caused by a non-linearity in the tuning of the laser this sweep can be done without detrimental broadening of the resulting spectral width. Method and apparatus of the invention avoid the aforementioned problems of the prior art and provide for exact data well keeping the measurement time low.

In a preferred embodiment of the invention there is performed a filtering of the Fourier transformed first signal with a high pass filter. It is further preferred to use a Hanning window as a shape for the high pass filter. Using this filter makes sure that a good elimination of not usable data is possible while still having important parts of data at the edges of the spectrum within the filter. Moreover, it is preferred to use a half of a Hanning window as a shape for the high-pass filter.

In a further preferred embodiment of the invention this high pass filter is adapted to the precision and the shape of the resulting spectrum of the signal corrected for a non-linearity of the laser by making an interferometric signal out of the corrected first phase signal, Fourier transforming the interferometric signal to get a spectral signal, determining a fraction of the maximum of the spectral signal, determining the abscissas of the intersections of the ordinate

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of the fraction which the curve of the spectral signal, determining the mean frequency f_{mean} as the average of the abscissas, band pass filtering the spectral signal with a band pass filter having its center at the mean frequency and having a width greater than the width of the frequency range. The inventive concept of adaptive filtering of the signal corrected for a non-linearity of the laser is doubling the inventive success since by eliminating the non-linearity the resulting peak of the Fourier transformed spectrum is very sharp so that it can be used a filter width which is very much smaller than known filter width in the prior art. As can be seen in the following detailed description of the invention and the respective drawings this inventive concept is resulting in very smooth curves showing the group delay.

In a preferred embodiment the inventive method comprises the following steps:

- splitting the first initial light beam into a first light beam and a second light beam,

- coupling the first light beam into the optical device under test,

- letting the second light beam travel a different path as the first light beam,

- superimposing the first and the second light beam to produce interference between the first light beam and the second light beam in a resulting first superimposed light beam,

- detecting as a first signal the power of the first superimposed light beam as a function of time when tuning the frequency of the coherent light beam from a minimum to a maximum of a given frequency range in a given time interval,

- splitting the second initial light beam in a third light beam and a fourth light beam,

- superimposing the third light beam and the fourth light beam after each light

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beam has traveled a different path, to produce interference between the third and the fourth light beam in a resulting second superimposed light beam,

- detecting as a second signal the power of the resulting second superimposed light beam as a function of time when tuning the frequency of the coherent light beam from a maximum to a minimum of a given frequency range in a given time interval,

- using the detected second signal for deriving a non-linearity information about a non-linearity in a tuning gradient of the frequency when tuning the frequency of the coherent light beam from the maximum to the minimum of the given frequency range, and

- using the non-linearity information for correcting effects on the first signal caused by the non-linearity to get a corrected first signal.

It is further preferred to flatten the first and/or the second signal before transforming it by approximating it with polynoms of at least second order.

Additionally it is preferred to tune the frequency with a tuning velocity of about 5 nm/s.

Additionally it is preferred to tune the frequency within a tuning interval of about c/1400nm to c/1600nm, c = velocity of light.

Additionally it is preferred to detect the power at a rate of about 100kHz.

The term "coherent" in this application means that the coherence length of the light beam is bigger than the difference of the path of the first and second and third and fourth light beams, respectively.

Other preferred embodiments are shown by the dependent claims.

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It is clear that the invention can be partly or entirely embodied or supported by one or more suitable software programs, which can be stored on or otherwise provided by any kind of data carrier, and which might be executed in or by any suitable data processing unit.

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BRIEF DESCRIPTION OF THE DRAWINGS

Other objects and many of the attendant advantages of the present invention will be readily appreciated and become better understood by reference to the following detailed description when considering in connection with the accompanied drawings. The components in the drawings are not necessarily to scale, emphasis instead being placed upon clearly illustrating the principles of the present invention. Features that are substantially or functionally equal or similar will be referred to with the same reference sign(s).

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Fig. 1   shows a chirped Bragg grating in a light path,

15   Fig. 2   shows an embodiment of the invention,

Fig. 3   shows a flow chart of the steps of an embodiment of the inventive method,

Fig. 4   shows an illustration of the step 100 of Fig. 3,

Fig. 5   shows an illustration of the step 106 of Fig.3,

20   Fig. 6   shows an illustration of the step 116 of Fig.3,

Fig. 7   shows two possible shapes for filters used in step 116 of Fig. 3,

Fig. 8   shows an illustration of the step 100 of Fig. 3, and

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Fig. 9   shows the group delay of the grating 12 of the example of Fig. 1.

DETAILED DESCRIPTION OF THE INVENTION

Referring now in greater detail to the drawings, Fig. 2 shows a schematic illustration of a preferred embodiment of an apparatus 10 for interferometric

5   determination of the group delay of a chirped Bragg grating 12 as a device under test. However, with the present invention it is possible to determine optical properties, e.g. the group delay ,of other devices, e.g. fibers or even air, also.

The apparatus 10 comprises a tunable laser 14 as a signal source for a

10   coherent laser beam 16, wherein the laser 14 is tunable between 1400 nm and at most 1600nm. The tunable laser 14 can be tuned continuously within a given frequency range in a given time interval. The laser beam 16 is coupled into a first beam splitter 18, which splits the coherent laser beam 16 in a first initial beam 20 and a second initial beam 22. The first initial beam 20 is coupled into

15   a second beam splitter 24. With the help of the second beam splitter 24 the first initial beam 20 is split into a first beam 26 and a second beam 28. The first beam 26 is traveling a different and longer path than the second beam 28, since the first beam 26 is coupled via a circulator 52 into the grating 12. Reflected by the grating 12 the first beam 26 is coupled via the circulator 52 in

20   the first path 32 of the first beam 26 again. Then the first beam 26 and the second beam 28 are united with a third beam splitter 34 by superimposing the first beam 26 and the second beam 28 to a first resulting beam 36. The first resulting beam 36 is coupled into a first photodiode 38. The photodiode 38 transmits its output as a first signal to an analog/digital-converter (ADC) 40

25   (which can be National Instruments AD-MIO-16DE-10). The ADC 40 is connected to a transient recorder 42 that can be a FAST TR1202 of Fast Comtec GmbH, Germany. The laser 14 can be a HP81680A of the applicant. The transient recorder 42 is coupled to a computer 44 as an evaluation unit of

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apparatus 10.

Second beam splitter 24, first beam 26, second beam 28 and third beam splitter 34 built up a first Mach-Zehnder-interferometer.

Exiting the first beam splitter 18 is a second initial beam 22. A fourth beam
5   splitter 46 splits the second initial beam 22 into a third beam 48 traveling a third path and a fourth beam 50. The third beam 48 is traveling a different and longer path than the fourth beam 50, wherein the path length of the third path differs from the path length of the fourth path by more than 1 m, preferred by less than 20 m, more preferred by less than 9 m, even more preferred between
10   7 and 9 m. This different path is symbolized by some loops 30. The third 48 and the fourth 50 beam are coupled into a fifth beam splitter 54 which units the third beam 48 and the fourth beam 50 to a superimposed second resulting beam 56. The second resulting beam 56 is then coupled into a second photodiode 58. The second photodiode 58 transmits its output as a second
15   signal to an analog/digital converter ADC 60, similar to the ADC 40. The ADC 60 is also connected to the transient receiver 42. Fourth beam splitter 46, third beam 48, fourth beam 50 and fifth beam splitter 54 built up a second Mach-Zehnder interferometer.

The system for determination of the group delay of the optical component 12
20   according to the present invention is based on the use of aforementioned Mach-Zehnder interferometers, both of which are supplied with light of the same light source 14. It is necessary to make the determination by interferometric measurements since the phase characteristics of the component 12 can not be detected directly.

25   The calculation of the properties based on the detected first and second signals is as follows:

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The laser 14 emits light 16 with field strength E dependent on the time t according to the following formula:

$$E_{Laser}(t) = E_0 \cdot \exp(j \cdot 2\pi \int_0^t v_{mom}(t')\,dt') \quad (3.1)$$

5   with $v_{mom} = \dfrac{1}{2\pi}\dfrac{d\phi}{dt}$ (3.2)

into the Mach-Zehnder interferometers, v_{mom} denoting the instantaneous frequency of the laser 14 .

In the embodiment of Fig. 2 the frequency of the laser 14 is changed in a linear
10   way so that the instantaneous frequency is

$$v_{mom}(t) = v_0 + \gamma \cdot t \quad (3.3)$$

To calculate as a second signal the power of the resulting signal 56 of the second Mach-Zehnder interferometer detected with the second detector 58,
15   one puts for the detected voltage U_{det} that the light of the longer path is delayed by $\tau = \Delta L \cdot n/c$ when arriving at the fifth beam splitter 54. n is the refractive index of the medium, ΔL is the difference of the path length and c is the velocity of light.

The detected voltage is determined by the two superimposed fields that travel
20   through the different paths of the interferometer:

$$U_{det}(t) \quad \sim |E_{Laser}(t) + E_{Laser}(t+\tau)|^2$$

$$= \left| \exp\left(j \cdot 2\pi \int_0^t v_{mom}(t')dt'\right) + \exp\left(j \cdot 2\pi \int_0^{t+\tau} v_{mom}(t')dt'\right) \right|^2$$

$$= 2\left[1 + \cos\left(2\pi \int_0^{t+\tau} v_{mom}(t')dt' - 2\pi \int_0^t v_{mom}(t')dt'\right) \right]$$

$$= 2\left[1 + \cos\left(2\pi \int_t^{t+\tau} v_{mom}(t')dt'\right) \right]$$

(3.4)

Only regarding of the changing part of this signal and putting $\Gamma = \int v(t) \, dt$ the result is:

$$U_{det}^{\sim}(t) = 2 \cdot \cos\left[2\pi \cdot (\Gamma_{mom}(t+\tau) - \Gamma_{mom}(t)) \right]. \quad (3.5)$$

Assuming $\tau \ll t$ and approximating with differential quotient one reaches:

$$U_{det}^{\sim}(t) = 2 \cdot \cos\left[2\pi \cdot \frac{\Gamma_{mom}(t+\tau) - \Gamma_{mom}(t)}{\tau} \cdot \tau \right] \quad (3.6)$$

$$= 2 \cdot \cos\left[2\pi \cdot \underbrace{\overline{V}_{mom}(t)}_{V_{mom} \, average \, over \, \tau} \cdot \tau \right] \quad (3.7)$$

The second signal of the second interferometer is therefore ideally, i.e. with a linear rise in frequency of the laser 14 over time a cosine, the frequency of which is determined by the length ΔL and the sweep velocity of the wavelength over time.

Because the laser 14 will produce not an ideal sweep over the given frequency range and will produce non linearities in this sweep one can expect a cosine which has a frequency which is changing over time. With this second signal it is possible according to the present invention to correct the first signal of the first interferometer for this non-linearities in therefore getting a resulting corrected first signal as if the laser 14 had performed a linear sweep.

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In the first interferometer in which the grating 12 is incorporated the group delay of this grating 12 is determined. For this purpose as the first signal the detected power of the first photodiode 38 is calculated in an analog way as explained above. However, in the following the first signal is only determined in its

5   dependency of the optical frequency υ, where as the time dependency is determined later on.

For the detected signal at the photodiode 38 the following formula applies:

$$U_{det}(v) \sim |\exp(-j \cdot \beta_{DUT}(v) \cdot L_{DUT}) + \exp(-j \cdot \beta(v) \cdot L)|^2$$
$$= 2[1 + \cos(\beta_{DUT}(v)L_{DUT} - \beta(v)L)] \qquad (3.8)$$

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Whereas β is the phase constant of the second path, whereas β_{DUT} is the phase constant of the first path through the grating 12. In this formula the fact is neglected that the grating 12 can have a reflection coefficient smaller than one. However, this has no importance for the changing part of the detected

15   signal 36. For this changing part it can be written with $\phi = \beta \cdot L$:

$$U_{det}^{\sim}(v) = 2 \cdot \cos(\phi_{DUT}(v) - \phi(v)) \qquad (3.9)$$

Hilbert-transforming this signal gives the analytic signal:

$$\hat{U}_{det}^{\sim}(v) = 2 \cdot \exp(\phi_{DUT}(v) - \phi(v)) \qquad (3.10)$$

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and the phase information:

$$\phi_{det}(v) = \phi_{DUT}(v) - \phi(v) \qquad (3.11)$$

To determine the time dependency of the signal when performing a frequency

25   sweep of the laser 14 and to find an evaluation rule for the group delay $\tau_{g,DUT}$, the functions ϕ and ϕ_{DUT} are now examined.

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For the phase of the path of the beam 28 without the grating 12 which path is assumed to be dispersion free it is possible to perform a Taylor approximation of the first order:

$$
\begin{aligned}
\phi(v) &= \phi(v_0) + \frac{d\phi}{dv}\bigg|v = v_0 \cdot (v - v_0) \\
&= \phi(v_0) + 2\pi \frac{d\phi}{d\omega}\bigg|\omega = \omega_0 \cdot (v - v_0) \\
&= \phi(v_0) + \tau(v_0) \cdot 2\pi(v - v_0) \\
&= \phi(v_0) + \frac{2L \cdot n(v_0)}{c} \cdot 2\pi(v - v_0)
\end{aligned}
\tag{3.12}
$$

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where L is the length of the path without the grating 12.

The same goes for the path 32 of the beam 26 with the grating 12:

$$
\begin{aligned}
\phi_{DUT}(v) &= \phi_{DUT}(v_0) + \frac{d\phi_{DUT}}{dv}\bigg|v = v_0 \cdot (v - v_0) \\
&+ \frac{1}{2}\frac{d^2\phi_{DUT}}{dv^2}\bigg|v = v_0 \cdot (v - v_0)^2 + \Psi,
\end{aligned}
\tag{3.13}
$$

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where ψ is denoting the residual error. Performing the same steps as in formula 3.12 one reaches the following formula:

$$
\begin{aligned}
\phi_{DUT}(v) &= \phi_{DUT}(v_0) + \tau_{DUT}(v_0) \cdot 2\pi(v - v_0) \\
&+ \frac{1}{2}\frac{d}{dv}(\tau_{DUT}(v))\bigg|v = v_0 \cdot 2\pi(v - v_0)^2 + \Psi
\end{aligned}
\tag{3.14}
$$

15

with

$$
\frac{d\tau_{DUT}}{dv} = \frac{d\tau_{DUT}}{d\lambda}\cdot\frac{d\lambda}{dv} = \frac{d\tau_{DUT}}{d\lambda}\cdot\frac{1}{\dfrac{d}{d\lambda}\left(\dfrac{c}{\lambda}\right)} = -\frac{d\tau}{d\lambda}\cdot\frac{\lambda^2}{c}.
\tag{3.15}
$$

20   By putting this formula into 3.11 this gives for the phase of the analytic signal:

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$$\phi_{det} = \phi_{DUT}(v_0) - \phi(v_0)$$
$$+ 2\pi\left[\tau_{DUT}(v_0) - \tau(v_0)\right]\cdot(v - v_0)$$
$$-\frac{d\tau}{d\lambda}\Big|\lambda_0\cdot\frac{\lambda^2}{c}\pi(v-v_0)^2 \qquad (3.16)$$
$$+\Psi$$

5        $$=\Delta\phi_0 \qquad (3.17)$$

$$+2\pi\Delta\tau_0\cdot(v-v_0) \qquad (3.18)$$

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$$-\frac{d\tau}{d\lambda}\Big|\lambda_0\cdot\frac{\lambda^2}{c}\pi(v-v_0)^2 \qquad (3.19)$$

$$+\Psi \qquad (3.20)$$

15   To determine the group delay of the grating 12 it is sufficient to differentiate the face ϕ_{det} with respect to ω. This gives:

$$\frac{d\phi_{det}}{d\omega} = \frac{1}{2\pi}\cdot\frac{d\phi_{det}}{dv} = \frac{1}{2\pi}\cdot\left[\frac{d\phi_{DUT}}{dv} - \frac{d\phi}{\underbrace{dv}_{const}}\right] \qquad (3.21)$$

20   This is the information searched for since only the behavior of the group delay with respect to the frequency is interesting, rather than an absolute value.

However, it is easy to extract from formulas 3.17 – 3.20 information about a deviation of this linear rise of the group delay with respect to the wavelength for
25   gratings with linear chirp. This goes by approximating the measured phase signal with a polynomial of second order and subtracting this polynomial of the phase signal. After this only the part 3.20 remains which represents the deviation of the group delay with respect a linear chirp grating.

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With the help of the diagram of Fig. 3 the main steps of an embodiment of the inventive method are described, whereas some more details of these steps follow later on:

In Fig. 3 ovals describe signals and boxes describe measures of the evaluation
5   unit 44. First of all the second or reference signal 56 and the first signal 36 are towed to the zero line in step 100 (see Fig. 4). The resulting signals 102 and 104 are now oscillating about the zero line. In step 106 any non-linearities in the sweep of the laser 14 are corrected with the help of the second signal 56 (see Fig. 5). Resulting is a signal 108 corrected for any non-linearities in the
10   laser sweep. With the help of the aforementioned formulas are evaluated in step 110 the group delay 112 and the linear deviation of linear chirp 114.

With the help of a filter step 116 it is possible to get filtered signals 118 and 120 of the group delay and of the deviation from the linear chirp of the grating 12.

In Fig. 4 the principle of the function 100 is graphically illustrated. A is the
15   amplitude of the signals 36 and 56. 152 denotes the reflection spectrum of the grating 12. 124 denotes the ideal center line of the spectrum 122.

The function 100 is illustrated in Fig. 8 and comprises the following steps:

- making the signal 36, 56 oscillating about a zero line by:

- determining the points of mean value of the signal 36, 56 by:

20   - determining the maximum and the minimum of the signal 36, 56 in a predetermined first range of time smaller than the total range of time,

- determining a mean value between the maximum and the minimum,

- determining the maximum and the minimum of the signal 36, 56 in a predetermined next range of time adjacent the already examined range of time,

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- determining a mean value between the maximum and the minimum,

- repeating the last two steps until the complete time interval is covered,

- choosing the predetermined range of time by:

- determining the average period of the oscillations of the signal 36, 56,

5   - choosing the size of the range so that more than two average periods fit in the chosen range of time.

Figure 5 illustrates a part of the function of step 106 to correct unwitting non-linearities in the sweep of the laser 14. The step 106 comprises the following steps (in the following the symbol "f" is used instead of "v"):

10   - deriving the non-linearity information by:

- transforming the second signal 56 to get a Fourier transformed second signal,

- eliminating the negative parts of the Fourier transformed second signal to get a non-negative Fourier transformed second signal,

- retransforming the non-negative Fourier transformed second signal to get an
15   analytic signal of the second signal 56,

- determining the phase of the analytic signal to get as a second phase signal the phase as a function of time of the second signal 56,

- using the second phase signal for determining as the non-linearity information the frequency as a function of time of the second signal 56,

20   - deriving a first phase signal by:

- transforming the first signal 36 to get a Fourier transformed first signal,

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- eliminating the negative parts of the Fourier transformed first signal to get a non-negative Fourier transformed first signal,

- retransforming the non-negative Fourier transformed first signal to get an analytic signal of the first signal 36,

5   - determining the phase of the analytic signal to get as a first phase signal as a function of time of the first signal,

- correcting the effects on the first signal 36 caused by the non-linearity by:

- using the non-linearity information to interpolate the first phase signal of the first signal 36 on a linear scale of frequencies to get a corrected first phase

10   signal,

For determining the non-linearity information the following steps are performed:

- determining the frequency $f(n)$ of the second signal 56 as a function of n discrete points of time, n= 1, ..., N, on the basis of the second phase signal by:

- determining the second phase signal $\phi(n)$ at the n points of time,

15   - determining the maximum ϕ_{max} of the second phase signal,

- using a predetermined maximum frequency f_{max} of the frequency range, a predetermined average tuning velocity during tuning the frequency and the maximum ϕ_{max} of the second phase signal to determine for each of the n points of time the frequency $f(n)$ according to the formula: $f(n)=[(f_{max}-f_{min})/\phi_{max}]\phi(n)$,

20   For the interpolating step, the following sub-steps are performed:

- getting the linear scale $f_{lin}(n)$ of frequencies by:

- using the predetermined maximum frequency f_{max} of the frequency range and

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the predetermined minimum frequency f_{min} of the frequency range to determine the linear scale $f_{lin}(n)$ of frequencies according to the formula: $f_{lin}(n)=[(f_{max}-f_{min})/(N-1)]n$,

- preferably sorting the absolute values of f(n) monotonically,

5 - using f(n) for interpolating the first phase signal of the first signal 36 on the linear scale of frequencies $f_{lin}(n)$. The interpolation can be done linearly or with the help of polynoms or splines.

The resulting signal 108 which is corrected for any non linearities in the sweep of the laser 14 is used in step 110 for evaluate the group delay 112 and the deviation 114 from the linear chirp of the grating 12. This evaluation is done with the help of the above-mentioned mathematical formulas and can comprises deriving a group delay 112 of the optical device under test 12 by differentiating the compensated first phase signal with respect to the frequency.

The method can further comprise the following steps:

15 - ignoring at the begin of the tuning a predetermined amount of values of the corrected first phase signal to eliminate teething troubles out of the corrected first signal 108,

- approximating the group delay 112 with polynoms of at least second order to get an approximated group delay,

20 - subtracting the approximated group delay from the group delay 112 to get a non-linear part of the group delay 112.

- using the non-linear parts of the group delay 112 to determine the mean signal power of a deviation from a linear of the group delay 112 of the device under test 12.

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- using the square coefficient of the polynomial to determine the mean gradient of the group delay 112.

Having evaluated the group delay 112 and the deviation 114 it is possible but not necessary to make a filter step 116 to filter the signals 112, 114. Actually, this filter step is preferably performed before the calculating of the group delay 112, as follows:

- adapting the filtering to the shape of the corrected first signal 108 by:

- a: making an interferometric signal out of the corrected first phase signal,

- b: Fourier transforming the interferometric signal to get a spectral signal,

- c: determining a fraction, preferred the half, of the maximum of the spectral signal,

- d: determining the abscissas of the intersections of the ordinate of the fraction with the curve of the spectral signal,

- e: determining the mean frequency f_{mean} as the average of the abscissas,

- f: band-pass filtering the spectral signal with a band-pass filter having its center at the mean frequency and having a width greater than the width of the frequency range,

- determining the width of the band-pass filter by:

- a: predetermining or estimating the maximal range GD_{range} of the group delay,

- b: determining the mean value GD_{mean} of the group delay according to the steps c-e of claim 26,

- c: calculating the filter width according to the formula: filter width=

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$f_{mean}(GD_{range}/GD_{mean})$,

- filtering with a filter having the shape of a Hanning window.

Resulting is a filtered group delay 118 and a filtered deviation 120.

Figure 6 shows the principle of determination of the center frequency f_{mean} of
the interferometric spectrum.

Figure 7 shows a graph showing different filter shapes for use with the inventive
method, a flat-top filter shape 130 and a Hanning form filter shape 140. Figure
9 shows the group delay 112 over wave length of a chirped Bragg grating 12,
which is a FBGDC-HP1A of the company JDS Fitel. It is clearly seen that the
graph of the group delay 112 is very smooth with respect to graphs determined
with methods of the prior art.

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